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Improved Strategy of Grid-Forming Virtual Synchronous Generator Based on Transient Damping

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ABSTRACT

The grid-forming virtual synchronous generator (GFVSG) not only employs a first-order low-pass filter for virtual inertia control but also introduces grid-connected active power (GCAP) dynamic oscillation issues, akin to those observed in traditional synchronous generators. In response to this, an improved strategy for lead-lag filter based GFVSG (LLF-GFVSG) is presented in this article. Firstly, the grid-connected circuit structure and control principle of typical GFVSG are described, and a closed-loop small-signal model for GCAP in GFVSG is established. The causes of GCAP dynamic oscillation of GFVSG under the disturbances of active power command as well as grid frequency are analyzed. On this basis, the LLF-GFVSG improvement strategy and its parameter design method are given. Finally, the efficiency of the proposed control strategy in damping GCAP dynamic oscillations under various disturbances is verified using MATLAB simulations and experimental comparison results.

KEYWORDS

Grid-forming virtual synchronous generator; first-order low-pass filter; lead-lag filter; small-signal model; parameter design

Nomenclature

Abbreviations

GFVSG	Grid-forming virtual synchronous generator
GCAP	Grid-connected active power
LLF-GFVSG	Lead-lag filter based GFVSG
RESs	Renewable energy sources
PECs	Power electronic converters
PEDPSs	Power electron dominated power systems
TSG	Traditional synchronous generator

Parameters and Constants

U_{dc}	DC side voltage
U_g	The amplitude of u_{gabc}



Z_{line}	Line impedance
C_f	Filter capacitor
θ	Output phase angle
u_{gabc}	Three-phase network voltage
i_{abc}	Grid-connected current
ω_0	Rated angular frequency
ω_g	Angular frequency of u_{gabc}
X	Equivalent reactance of Z_{line}
Δ	The amount of perturbation/fluctuation

Variables

P_e	Grid-connected active power
P_{ref}	Reference power
J	Virtual inertia
D	Virtual damping
ω	Output angular frequency
Q_e	Grid-connected reactive power
Q_{ref}	Reference reactive power
E	Output voltage amplitude
E_0	Nominal voltage
k_q	Proportional gain
K	Synchronous voltage coefficient
δ	Phase angle
E_0	Nominal voltage
ω_n	Undamped oscillation frequency of second-order system
ξ	Damping ratio
ΔP_{e0}	The GCAP steady state deviation
K_p	The forward compensation coefficient
K_d	The feedforward compensation coefficient

1 Introduction

In recent years, nations worldwide have confronted varying degrees of energy pressures: firstly, the challenge of securing adequate and reliable energy supplies at sustainable prices over the long term, and secondly, the profound and irreversible environmental damage resulting from excessive energy consumption. It is in this energy dilemma that a range of renewable energy sources (RESs) such as hydro, wind and solar, which have high resource potential and low environmental pollution, have begun to develop rapidly [1]. RES inevitably requires power electronic converters (PECs) as its grid-connected interface, which promotes the transformation of traditional power systems to power electron dominated power systems (PEDPSs) [2,3]. While PECs bring controllability as well as flexibility to PEDPS, they also bring some drawbacks, unfortunately, such as a significant reduction in inertia and damping levels, which leads to several stability problems, e.g., dynamic oscillations of active power and output frequency [4]. In order to eliminate the dynamic oscillations of PEDPS active power and output frequency caused by low inertia and weak damping, the grid-forming virtual synchronous generator (GFVSG) is proposed to provide sufficient inertia and damping support for PEDPS [5–7]. GFVSG simulates the rotor motion equation of a traditional synchronous generator (TSG), that is, it uses the first-order low-pass filter containing virtual inertia and virtual damping parameters

to realize virtual inertia and virtual damping control, which helps to improve the PEDPS frequency stability [8,9].

It can be pointed out that the GFVSG, while realizing the virtual inertia control, makes its closed-loop control system of grid-connected active power (GCAP) into a typical second-order oscillation system, which is prone to lead to the dynamic oscillation phenomenon of GCAP similar to that of the TSG under the perturbations of active power command, grid frequency and so on [10,11]. The large fluctuating current existing in the GCAP dynamic oscillation of GFVSG is easy to cause the PEC with weak overload capability to shutdown due to the over-current protection or even hardware equipment to burn down, which reduces the reliability of GFVSG grid-connected operation system [12]. At present, the control methods applied to suppress or eliminate the dynamic oscillations of GCAP and output frequency of GFVSG mainly include adaptive parameter tuning scheme [13–16], dynamic feedback compensation method [17–20] and dynamic feedforward compensation approach [21–24].

Among them, the adaptive parameter tuning scheme optimizes the GCAP dynamic response performance of GFVSG by leveraging the fact that parameters such as virtual inertia, virtual damping, or virtual impedance of GFVSG can be flexibly adapted online without being restricted by the physical conditions. In [13], the virtual inertia parameter was adaptively adjusted according to the sign of the product of the GFVSG output angular frequency and the rate of change of the angular frequency to suppress the GCAP dynamic oscillation. In [14,15], an adaptive parameter tuning scheme by simultaneously and adaptively adjusting both the virtual inertia parameter and the virtual damping parameter was used to further achieve the optimization of the GCAP dynamic response performance of GFVSG. Reference [16], an adaptive virtual impedance adjustment method, was applied to the GFVSG grid-connected system to enhance the system's ability to damp GCAP dynamic oscillations. It is noteworthy that the adaptive parameter tuning methods discussed in references [13–16] necessitate online modifications to the key parameters of GFVSG, thereby introducing nonlinear variation characteristics into the GFVSG grid-connected system. This complexity escalates the challenges of system parameter tuning and heightens the risk of operational instability.

Different from the adaptive parameter tuning scheme, the dynamic feedback compensation method and the dynamic feedforward compensation approach improve the GCAP dynamic response characteristics of GFVSG by reconfiguring the control loop of the GCAP closed-loop control system of GFVSG under the premise of ensuring that the key parameters of GFVSG remain unchanged. In [17,18], the differential feedback links based on the GCAP and the output angular frequency were introduced into the GFVSG control loop, respectively. However, the digital implementation of the differential algorithms contained in [17,18] would bring about the problems of high frequency harmonic interference. In [19], the GCAP proportional feedback link was replaced by using a proportional feedback link based on a lead-lag filter, which avoids differential manipulations but increases the control order of the GCAP closed control system. In [20], a first-order lag link based on the GCAP was added into the GFVSG feedback control loop, which has a similar control effect as in [19], but increases the difficulty of parameter design. It should be noted that the dynamic feedback compensation methods used in [17–20] can only produce the effect after the deviation of the feedback variables, and there is a certain passivity in the control implementation.

Different from the dynamic feedback compensation method, the dynamic feedforward compensation approach has some initiative in control effect. In [21], the GCAP differential feedforward link was added into the control loop of GFVSG, but the GCAP differential operation would inevitably introduce high frequency harmonic signals. In [22], an angular frequency feedforward link based

on a lead-lag filter was added into the GFVSG control loop, which does not require differential operation, but increases the control order of the system as well as the difficulty of parameter design. In [23], a feedforward link based on the active power command was inserted into the control loop of GFVSG, which has the benefit of intuitive parameter design, but the parameter design depends on the line impedance parameters of the system. In [24], a phase-based dynamic feedforward compensation was proposed to optimize the GCAP dynamic response performance of GFVSG. The feedforward parameter design in [24] does not depend on the line impedance parameters, but the system suffers from a weakened ability to suppress high frequency interference signals.

To overcome the shortcomings of the above the dynamic feedback compensation methods and the dynamic feedforward compensation approaches, such as requiring differential operation, increasing control order and complex parameter design, an improved GFVSG based on lead-lag filter (LLF-GFVSG) is proposed in this paper. The remainder of this paper is organized as follows: Section 2 introduces the principle of GFVSG and analyzes the GCAP response characteristics. Section 3 proposes the LLF-GFVSG control strategy and provides a detailed parameter design method. Section 4 presents results of both simulation and experimental comparison. Finally, Section 5 concludes this paper.

2 Principle of GFVSG and Its GCAP Response Characteristics

2.1 Grid-Connected Structure of GFVSG and Its Control Principle

The GFVSG grid-connected circuit structure and its control principle are shown in Fig. 1 [11].

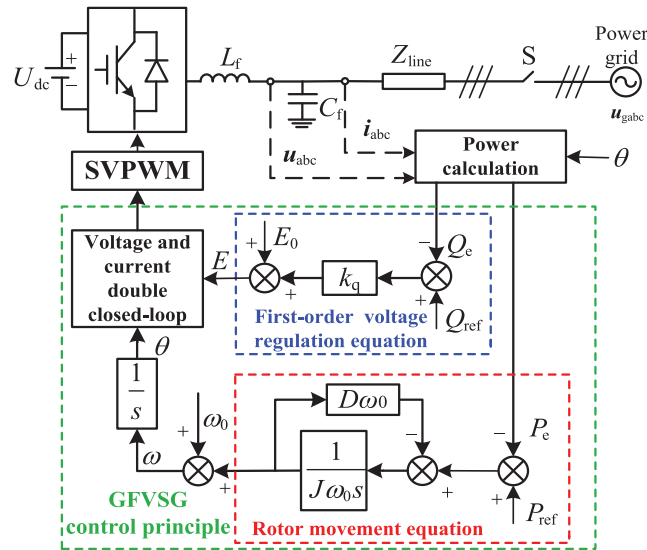


Figure 1: Grid-connected circuit structure and control principle of GFVSG

In Fig. 1, U_{dc} denotes DC side voltage and Z_{line} represents line impedance. L_f and C_f are filter inductance and filter capacitor; θ denotes output phase angle; u_{gabc} and S are three-phase network voltage and grid-connected switch; u_{abc} denotes three-phase output voltage and i_{abc} represents grid-connected current. The rotor equations of motion of the GFVSG and its primary regulating voltage equation can be expressed sequentially as follows:

$$P_{\text{ref}} - P_e - D\omega_0 (\omega - \omega_0) = J\omega_0 \frac{d\omega}{dt} = J\omega_0 \frac{d(\omega - \omega_0)}{dt} \quad (1)$$

$$E = E_0 + k_q (Q_{\text{ref}} - Q_e) \quad (2)$$

where P_e and P_{ref} are GCAP and its reference power, respectively, and J, D are virtual inertia and virtual damping; ω and ω_0 are output angular frequency and the fundamental; Q_e denotes grid-connected reactive power and Q_{ref} represents reference reactive power; E and E_0 are output voltage amplitude and nominal voltage, respectively, and k_q is a proportional gain.

It is worth noting that the primary focus of this paper lies in addressing the GCAP dynamic response optimization issue of GFVSG, considering the GCAP and the grid-connected reactive power can be decoupled under the condition that Z_{line} is inductive and disregarding the impact of the voltage and current bottom double-closed-loop on the power outer-loop with a high control bandwidth. Thus, the relevant content concerning the reactive power control and the bottom double-loop control will not be reiterated [21].

2.2 Small-Signal Model and Characteristic Analysis of GFVSG

As illustrated in Fig. 1 and combined with the line power transfer theory, the P_e of GFVSG can be approximated by Eq. (3) [25].

$$P_e = \frac{3U_g E}{2X} \sin \delta \approx K\delta = K \frac{\omega - \omega_g}{s} \quad (3)$$

where U_g is the amplitude of u_{gabc} and ω_g is the angular frequency of u_{gabc} . X denotes the equivalent reactance of Z_{line} ; the synchronizing voltage coefficient is first formulated as $K = 1.5U_g E/X$; δ and s are phase angle and laplacian. As shown in Fig. 2, the GCAP closed-loop small-signal control model of GFVSG can be derived from Eqs. (1) and (3). In Fig. 2, “ Δ ” represents the amount of perturbation.

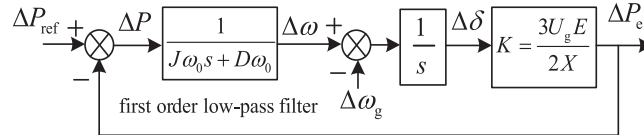


Figure 2: The GCAP closed-loop small-signal control model of GFVSG

From Fig. 2, it can be seen that while the GFVSG realizes the virtual inertia control by using the first-order low-pass filter $1/(J\omega_0s + D\omega_0)$ consisting of J and D , its P_e and ω are affected by ΔP_{ref} and $\Delta\omega_g$ disturbances. The transfer functions of ΔP_{ref} to ΔP_e , $\Delta\omega_g$ to ΔP_e , ΔP_{ref} to $\Delta\omega$ and $\Delta\omega_g$ to $\Delta\omega$ for GFVSG are derived as follows:

$$\begin{cases} G_1(s) = \left. \frac{\Delta P_e}{\Delta P_{\text{ref}}} \right|_{\Delta\omega_g=0} = \frac{K}{J\omega_0s^2 + D\omega_0s + K} \\ G_2(s) = \left. \frac{\Delta P_e}{\Delta\omega_g} \right|_{\Delta P_{\text{ref}}=0} = -\frac{(J\omega_0s + D\omega_0)K}{J\omega_0s^2 + D\omega_0s + K} \end{cases} \quad (4)$$

$$\begin{cases} G_3(s) = \left. \frac{\Delta\omega}{\Delta P_{\text{ref}}} \right|_{\Delta\omega_g=0} = \frac{s}{J\omega_0s^2 + D\omega_0s + K} \\ G_4(s) = \left. \frac{\Delta\omega}{\Delta\omega_g} \right|_{\Delta P_{\text{ref}}=0} = \frac{K}{J\omega_0s^2 + D\omega_0s + K} \end{cases} \quad (5)$$

According to Eqs. (4) and (5), it is easy to find that the GCAP closed-loop control system of GFVSG is a second-order oscillation system, and the corresponding natural oscillation angular frequency ω_n and its damping ratio ξ of the system can be expressed as:

$$\omega_n = \sqrt{\frac{K}{J\omega_0}}, \xi = \frac{D\omega_0}{2} \sqrt{\frac{1}{KJ\omega_0}} \quad (6)$$

It can also be obtained from Eq. (4) that the GCAP steady state deviation ΔP_{e0} ($\Delta P_{e0} = \Delta P_e - \Delta P_{ref}$) of GFVSG is shown in Eq. (7).

$$\Delta P_{e0} = \lim_{s \rightarrow 0} G_2(s) \Delta \omega_g = -D\omega_0 \Delta \omega_g \quad (7)$$

By applying Eqs. (6) and (7), it can be found that the value of J affects the ξ and ω_n of the GFVSG grid-connected system at the same time, i.e., the larger J is, the smaller the corresponding ξ and ω_n is, and the more pronounced the dynamic oscillation and the longer the dynamic response time of its GCAP under the disturbance of ΔP_{ref} and $\Delta \omega_g$. The value of D also affects ξ and ΔP_{e0} of the GFVSG grid-connected system; that is, the larger the value of D , the larger the corresponding ξ and ΔP_{e0} are, and the stronger the damping ability of GCAP dynamic oscillation, but the larger the GCAP steady state deviation of P_e when ω_g deviates from ω_0 . Consequently, for the GFVSG grid-connected system, the values of J and D dominated the dynamic and steady state response performance of its GCAP. In other words, the values of J and D could only be reasonably determined on the premise of weighing the GCAP dynamic and steady state response performance of GFVSG, resulting in certain limitations in the optimization of its GCAP response performance.

3 LLF-GFVSG Improvement Strategy

To overcome the inherent limitations of the GFVSG, arising from its dependency on a first-order low-pass filter for optimizing both dynamic and steady-state response performance of the GCAP, as depicted in Fig. 1, we have substituted the virtual inertia control component with a lead-lag filter. This results in the development and proposition of an enhanced LLF-GFVSG control strategy. This innovative approach not only provides the advantages of active control but also obviates the necessity for differential operations, preserves a second-order control framework, and simplifies the parameter design process. Additionally, it provides the parameter design methodology for LLF-GFVSG.

3.1 Control Principle of LLF-GFVSG

Fig. 3a shows the GCAP closed-loop small-signal control model of LLF-GFVSG. In Fig. 3a, K_p and K_d are the forward compensation coefficient and feedforward compensation coefficient of LLF-GFVSG, respectively. Further, the GCAP closed-loop small-signal equivalent control model of LLF-GFVSG can be obtained by equivalent transformation of the control block diagram in Fig. 3a, as shown in Fig. 3b. Since LLF composed of J , D , K_p and K_d is included in Fig. 3b, the improved strategy is referred to as LLF-GFVSG for short in this text.

According to Fig. 3a,b, the transmission functions of ΔP_{ref} to ΔP_e , $\Delta \omega_g$ to ΔP_e , ΔP_{ref} to $\Delta \omega$ and $\Delta \omega_g$ to $\Delta \omega$ for LLF-GFVSG can be expressed as:

$$\begin{cases} G_{11}(s) = \frac{K(K_d J\omega_0 s + K_p)}{J\omega_0 s^2 + (D\omega_0 + KK_d J\omega_0)s + KK_p} \\ G_{21}(s) = -\frac{(J\omega_0 s + D\omega_0)K}{J\omega_0 s^2 + (D\omega_0 + KK_d J\omega_0)s + KK_p} \end{cases} \quad (8)$$

$$\begin{cases} G_{31}(s) = \frac{s(K_d J\omega_0 s + K_p)}{J\omega_0 s^2 + (D\omega_0 + KK_d J\omega_0)s + KK_p} \\ G_{41}(s) = \frac{K(K_d J\omega_0 s + K_p)}{J\omega_0 s^2 + (D\omega_0 + KK_d J\omega_0)s + KK_p} \end{cases} \quad (9)$$

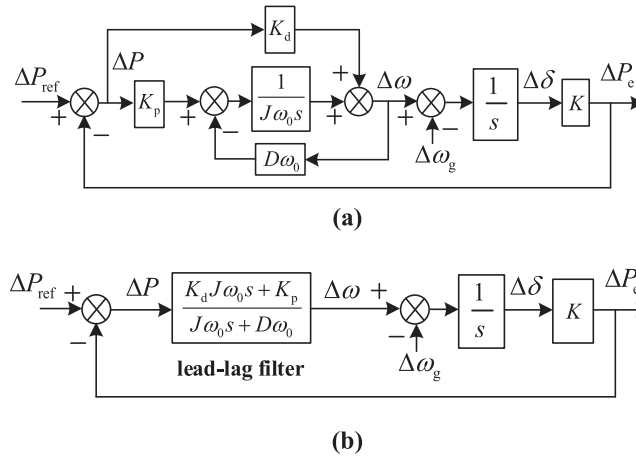


Figure 3: The GCAP closed-loop small-signal control model of LLF-GFVSG. (a) Control model of LLF-GFVSG. (b) Equivalent control model of LLF-GFVSG

It was comparing Eqs. (8) with (4) and Eqs. (9) with (5), respectively, it can be found that, compared with the GFVSG, the LLF-GFVSG, under the premise of ensuring that the order of its GCAP closed-loop control system remains unchanged and is still a second-order system, transforms the natural oscillating angular frequency of its grid-connected system, ω_{n1} , and its damping ratio, ξ_1 , in turn, as follows:

$$\omega_{n1} = \sqrt{\frac{KK_p}{J\omega_0}}, \xi_1 = \frac{D\omega_0 + KK_d J\omega_0}{2} \sqrt{\frac{1}{KK_p J\omega_0}} \quad (10)$$

From Eq. (8), the GCAP steady state deviation ΔP_{e01} of LLF-GFVSG is obtained, as shown in Eq. (11).

$$\Delta P_{e01} = \lim_{s \rightarrow 0} G_{21}(s) \Delta\omega_g = - (D\omega_0/K_p) \Delta\omega_g \quad (11)$$

By comparing Eqs. (10) with (6) and Eqs. (11) with (7), respectively, it can be found that the value of K_p affects ω_{n1} , ξ_1 and ΔP_{e01} of the LLF-GFVSG grid-connected system at the same time, in order to prioritize to ensure that the GCAP of LLF-GFVSG does not generate steady state deviation under the condition of ω_g deviation from ω_0 , i.e., to prioritize to satisfy the condition of $\Delta P_{e01} = \Delta P_{e0}$, the paper needs to set $K_p = 1$, then we have $\omega_{n1} = \omega_n$, which then ensures that the grid-connected systems of LLF-GFVSG and GFVSG have the same natural oscillation angular frequency, which provides a fair

condition for comparing the GCAP dynamic and steady state response performances of LLF-GFVSG and GFVSG, and at the same time simplifies the process of parameter design of LLF-GFVSG.

3.2 Parameter Design of LLF-GFVSG

With the theoretical analysis in the previous subsection and the setting of $K_p = 1$ ($\Delta P_{e01} = \Delta P_{e0}$ with $\omega_{n1} = \omega_n$), Eq. (10) can be equated as:

$$\omega_{n1} = \sqrt{\frac{K}{J\omega_0}}, \xi_1 = \frac{D\omega_0 + KK_d J\omega_0}{2} \sqrt{\frac{1}{KJ\omega_0}} \quad (12)$$

Comparing Eqs. (12) and (6), it is easy to see that LLF-GFVSG has one more K_d control degree of freedom than GFVSG, that is, the former can optimize its GCAP dynamic response performance directly by selecting a reasonable K_d value. It can also be seen from Eq. (8) that $G_{11}(s)$ of LLF-GFVSG is a second-order control system containing negative real zeros under the condition of $K_d > 0$. With reference to the parameter design method given in [25], $G_{11}(s)$ can be equivalent transformed into:

$$G_{11}(s) = \frac{K_d J\omega_0 s \omega_{n1}^2}{s^2 + 2\xi_1 \omega_{n1} s + \omega_{n1}^2} + \frac{\omega_{n1}^2}{s^2 + 2\xi_1 \omega_{n1} s + \omega_{n1}^2} \quad (13)$$

As can be seen from Eq. (13), the GCAP response of LLF-GFVSG mainly includes two parts: differential dynamic response and zero-free typical second-order system response. If K_d is larger, the negative real zero $z_0 = -1/(K_d J\omega_0)$ of $G_{11}(s)$ will be closer to the origin. The effect of differential dynamic response on the GCAP dynamic response of LLF-GFVSG is more obvious. At the same time, to suppress the GCAP dynamic oscillation of LLF-GFVSG, it is necessary to adjust the value of K_d to make $\xi_1 \geq 1$, and when $\xi_1 \geq 1$, it can further ensure that there is no power overshoot in the dynamic response process of the zero-free typical second-order system in $G_{11}(s)$.

On the one hand, under the condition that $\xi_1 \geq 1$, it can be obtained by Eq. (12):

$$\xi_1 \geq 1 \Rightarrow K_d \geq \frac{2\sqrt{KJ\omega_0} - D\omega_0}{KJ\omega_0} \quad (14)$$

On the other hand, under the condition that the value of K_d satisfies Eq. (14), regardless of the fact that the two negative real poles s_1 and s_2 of the second-order system with negative real zeros are different or the same, if z_0 is at the right end of s_1 and s_2 , i.e., if z_0 is closer to the origin and corresponds to the K_d obtaining a larger value. The differential dynamic response part of $G_{11}(s)$ will have a significant impact on the GCAP dynamic response of LLF-GFVSG and even introduce power overshooting [26]. Given this, it is recommended to select the z_0 of $G_{11}(s)$ in the range s_1 to s_2 , that is:

$$\underbrace{\left(-\xi_1 - \sqrt{\xi_1^2 - 1}\right) \omega_{n1}}_{s_1} \leq \underbrace{\frac{-1}{K_d J\omega_0}}_{z_0} \leq \underbrace{\left(-\xi_1 + \sqrt{\xi_1^2 - 1}\right) \omega_{n1}}_{s_2} \quad (15)$$

According to Eqs. (12), (14) and (15) as well as the main parameters of the GFVSG, the rationalization of the K_d parameters of the additional control degrees of freedom of the LLF-GFVSG can be accomplished.

4 Simulation and Experimental Test Results

4.1 Comparative Analysis of Simulation Results

To verify the correctness and effectiveness of the LLF-GFVSG improvement strategy and its parameter design method, the 100kVA-GFVSG grid-connected system simulation model as shown in Fig. 1, is built by means of MATLAB simulation program. In the simulation process, $P_{ref} = 0$ kW, $J = 6$ kg·m², $D = 50.66$, $K_p = 1$ for 100kVA-GFVSG is set. Other main simulation parameters for 100kVA-GFVSG are shown in Table 1 [11].

Table 1: Key parameters of 100kVA-GFVSG

Parameter	Value	Parameter	Value
Isolation trans-former	270 V:400 V	DC side voltage/ U_{dc}	650 V
Nominal voltage	380 V	Proportional gain/ k_q	1.4×10^{-4}
Fundamental frequency	50 Hz	Voltage loop proportional coefficient	750
Sample frequency	5 kHz	Voltage loop integral coefficient	50.6
Filter inductance/ L_f	0.56 mH	Current loop proportional coefficient	0.93
Filter capacitor/ C_f	270 uF	Equivalent reactance/ X	0.1Ω

At the same time, according to the main parameters given in Table 1 can be calculated to get $K = 1,452,000$, $\omega_n = \omega_{n1} = 27.7$ rad/s, $\xi = 0.15 < 1$, and need to set $K_d \geq 3.24 \times 10^{-5}$ to ensure that $\xi_1 \geq 1$, in this case, $K_d = 5.3 \times 10^{-5}$ is selected, and thus there is $\xi_1 = 1.52 > 1$, $z_0 = -10$, the $s_1 = -75$, $s_2 = -10$. It is worth pointing out that in Fig. 3a, K_d is multiplied by ΔP and then feedforward compensated to $\Delta\omega$, and for the 100kVA-GFVSG grid-connected system, usually ΔP (order of magnitude 105) $\gg \Delta\omega$ (unit of the order of magnitude), so that if the two reach a comparable order of magnitude, it is necessary for K_d to take a magnitude of 10^{-5} values.

By inputting the aforementioned parameters into Eqs. (8) and (9), Bode plots comparing the frequency response characteristics of $\Delta P_e/\Delta P_{ref}$, $\Delta P_e/\Delta\omega_g$, $\Delta\omega/\Delta P_{ref}$ and $\Delta\omega/\Delta\omega_g$ for LLF-GFVSG, GFVSG ($D = 50.66$ J/rad) and GFVSG ($D = 335.16$ J/rad) are presented in Fig. 4.

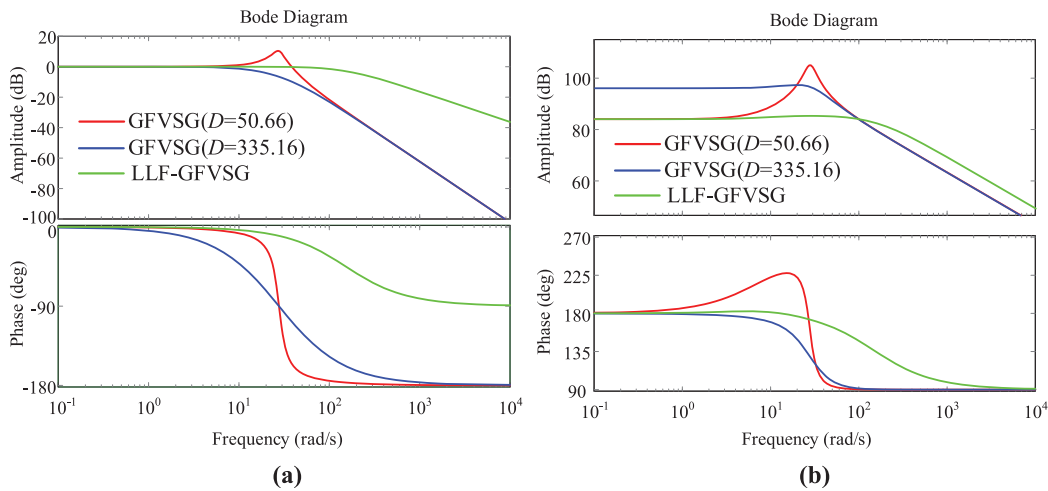


Figure 4: (Continued)

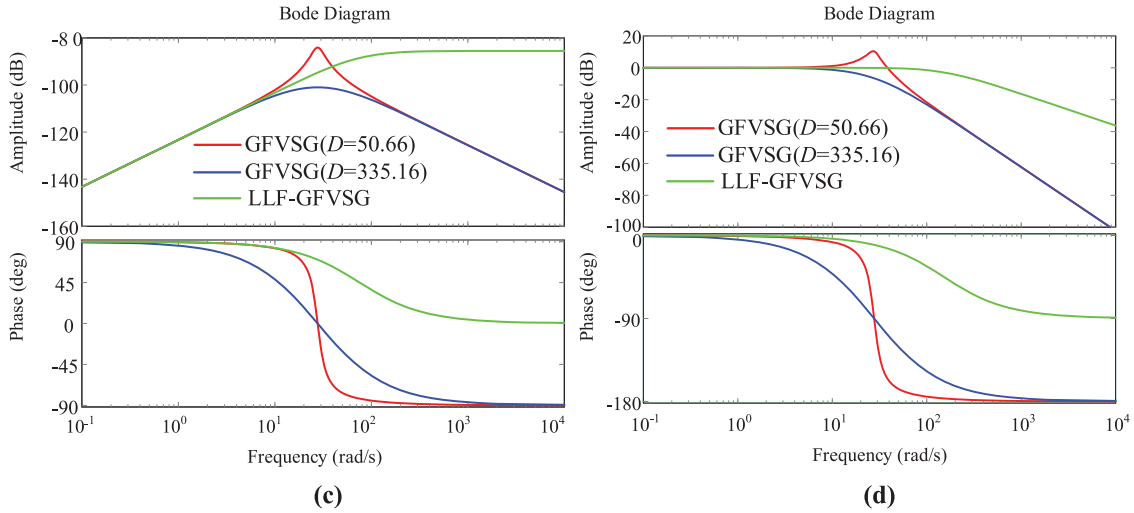


Figure 4: Frequency response comparison of LLF-GFVSG and GFVSG. (a) $\Delta P_e/\Delta P_{ref}$. (b) $\Delta P_e/\Delta \omega_g$. (c) $\Delta \omega/\Delta P_{ref}$. (d) $\Delta \omega/\Delta \omega_g$

Fig. 5 shows the simulation comparison results of P_e vs. output frequency f for LLF-GFVSG, GFVSG ($D = 50.66$) and GFVSG ($D = 335.16$), respectively, in response to the dynamic process of the P_{ref} stepping from 20 to 60 kW and the grid frequency f_g stepping from 50 to 49.95 Hz.

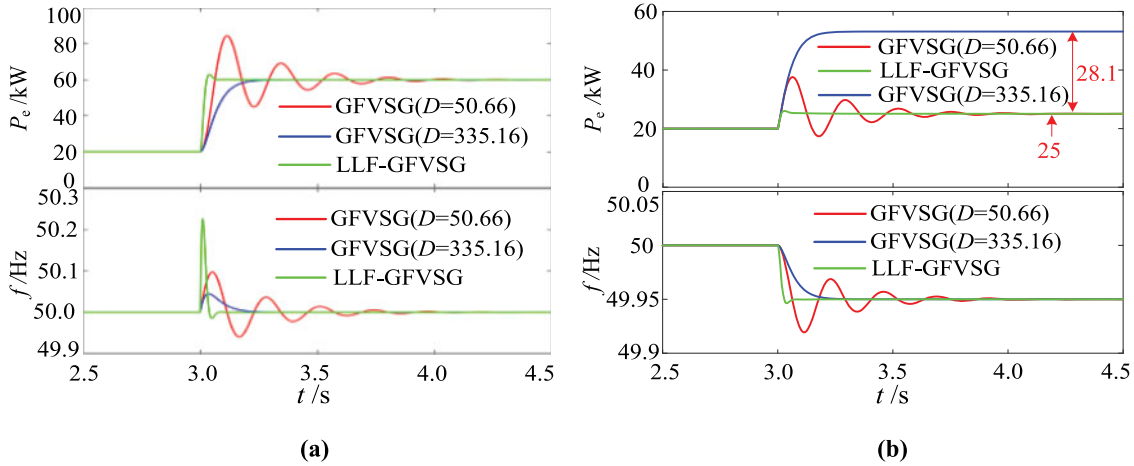


Figure 5: Comparative simulation results under P_{ref} step and f_g step. (a) P_{ref} step. (b) f_g step

It is not difficult to see from Figs. 4 and 5 that when $D = 50.66$, GFVSG is $\xi = 0.15$ of the underdamped system. The four bode diagrams representing GFVSG ($D = 50.66$) shown in Fig. 4 all have a resonant peak before the cutoff frequency. Therefore, its P_e and f under both the P_{ref} and f_g disturbances appear dynamic oscillations. When $D = 335.16$ is increased, GFVSG is an overdamped system $\xi = 1.006$, and the four bode diagrams representing GFVSG ($D = 335.16$) shown in Fig. 4 do not have a resonance peak before the cutoff frequency, so P_e and f do not oscillate dynamically under P_{ref} and f_g disturbances. However, its P_e introduces active steady state deviation at $f_g = 49.95$ Hz ($\Delta P_{e0} = 28.1$ kW); while the P_e and f of LLF-GFVSG do not show dynamic oscillations under the

two perturbations of P_{ref} and f_g , and its P_e does not have active steady state deviation at $f_g = 49.95$ Hz ($\Delta P_{e01} = 0$) as shown in Fig. 5.

Fig. 6 presents the simulation results of the P_e and f during the implementation of the FOVSG control method in reference [27], as well as the active reference command P_{ref} of LLF-GFVSG transitioning from 20 to 60 kW and power grid frequency f_g varying from 50 to 49.95 Hz under the identical conditions. As illustrated in Fig. 6, the LLF-GFVSG and FOVSG exhibit comparable dynamic performance in terms of active power P_e under disturbances in P_{ref} and f_g . However, the output frequency overshoot observed in FOVSG is significantly higher than that in LLF-GFVSG. During the f_g step change, the output frequency response f of LLF-GFVSG is observed to be slower than that of FOVSG, suggesting a superior inertia response in LLF-GFVSG compared to FOVSG.

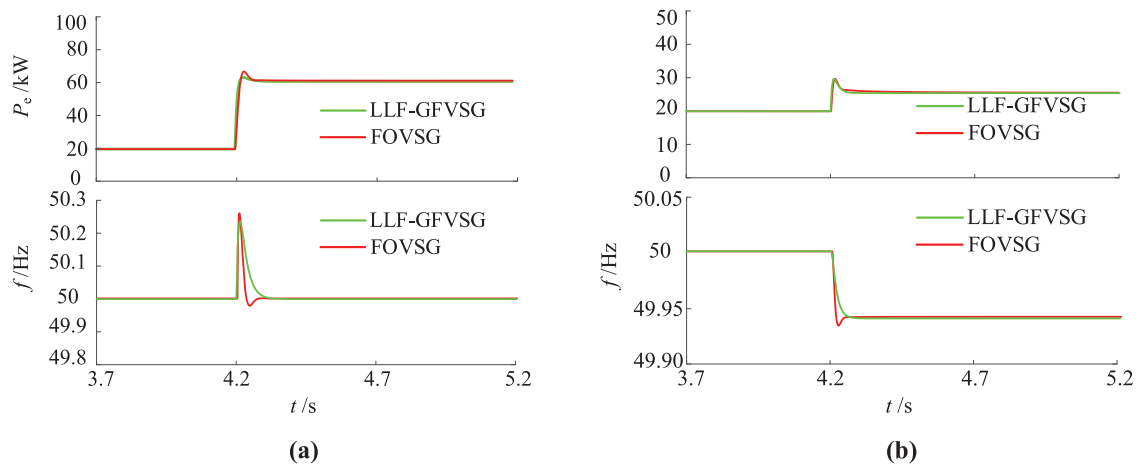


Figure 6: Comparison of simulation results between LLF-GFVSG and FOVSG. (a) P_{ref} step. (b) f_g step

In summary, the LLF-GFVSG can effectively solve the problem that it is challenging to balance the GCAP dynamic response performance and its steady state response performance of GFVSG under both the disturbances of P_{ref} and f_g , i.e., the LLF-GFVSG can ensure that its P_e has both good dynamic and steady state response performance under the two disturbances of P_{ref} and f_g . It should be noted that since the LLF-GFVSG directly utilizes ΔP to feedforward compensate $\Delta\omega$, and there is a large deviation between P_{ref} and P_e at the beginning of the disturbance, i.e., the dynamic feedforward compensation to $\Delta\omega$ is larger, so while improving the dynamic response speed of P_e , it is also easy to cause a large overshoot amplitude of f under the disturbance of P_{ref} .

4.2 Experimental Comparison Results and Analysis

In order to further verify the efficacy and superiority of the described LLF-GFVSG over the GFVSG in optimizing its GCAP dynamic response performance, experimental comparative verification is carried out on the energy storage microgrid system test platform presented in Fig. 7. The testing platform mainly includes two 100kVA-GFVSGs, two 100kVA bi-directional controllable rectifiers that provide a stable DC voltage and bi-directional energy supply for the 100kVA-GFVSGs (which can be used as a storage battery simulator), and a set of 250 kW resistive controllable loads and the distribution network [25].



Figure 7: Energy storage microgrid system platform

During the experimental test, $P_{\text{ref}} = 20 \text{ kW}$, $J = 6 \text{ kg}\cdot\text{m}^2$, $D = 50.66$, $K_p = 1$, $K_d = 5.3 \times 10^{-5}$ were set for 100kVA-GFVSG. Other main experimental parameters were consistent with [Table 1](#) and the simulation parameters. [Fig. 8a](#) shows the experimental results under a power grid voltage step change of -1% . [Fig. 8b](#) shows the experimental outcomes for Q_{ref} transitioning from 0 to 30 kvar. As inferred from [Eq. \(2\)](#), the primary voltage regulation equation represents a first-order system. Regardless of whether the voltage step is -1% or the Q_{ref} steps from 0 to 30 kvar, the change in Q_e remains non-oscillatory. Nevertheless, [Fig. 8](#) demonstrates that the LLF-GFVSG can effectively suppress dynamic oscillations in active power caused by grid voltage and Q_{ref} disturbances. Consequently, this study focuses exclusively on the changes in the grid-connected active power P_e and output frequency f .

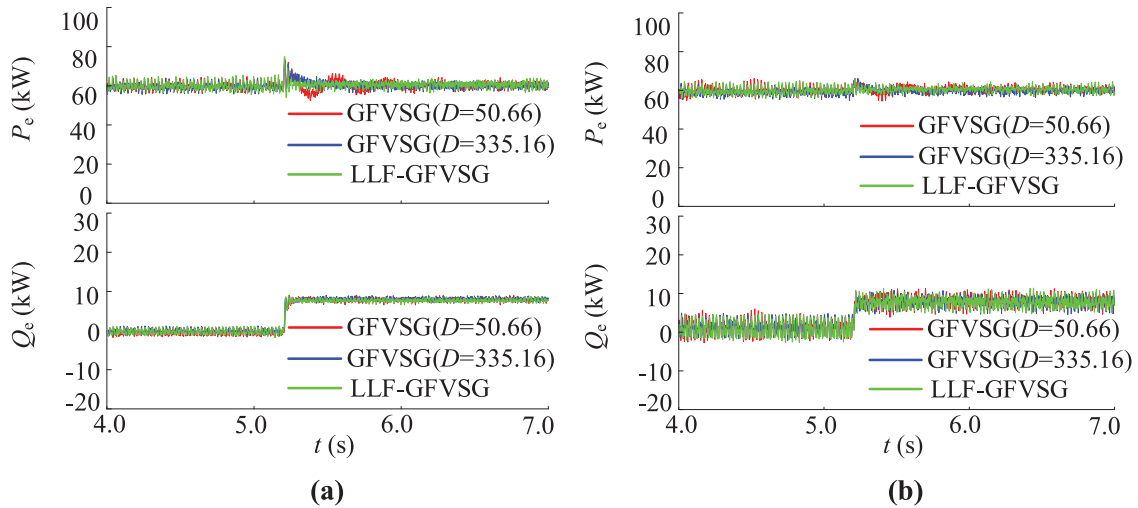


Figure 8: Comparative experimental results under (a) power grid voltage step -1% and (b) Q_{ref} step

[Figs. 9a](#) and [10a](#) illustrate the experimental results for the P_e and f under the condition of $X = 0.1 \Omega$, respectively. The comparison includes LLF-GFVSG, GFVSG ($D = 50.66$), and GFVSG ($D = 335.16$) during the transitions of P_{ref} from 20 to 60 kW and f_g from 50 to 49.95 Hz. Conversely, [Figs. 9b](#) and [10b](#) illustrate the experimental results for the P_e and f under the condition of $X = 0.05 \Omega$, respectively. The comparison includes LLF-GFVSG, GFVSG ($D = 50.66$), and GFVSG ($D = 335.16$) during the transitions of P_{ref} from 20 to 60 kW and f_g from 50 to 49.95 Hz.

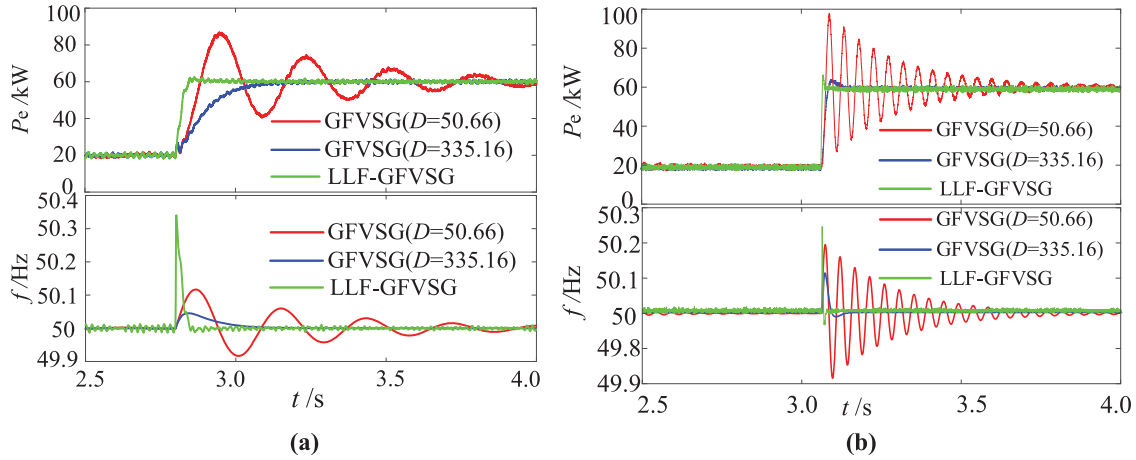


Figure 9: Comparative experimental results under P_{ref} step. (a) $X = 0.1 \Omega$. (b) $X = 0.05 \Omega$

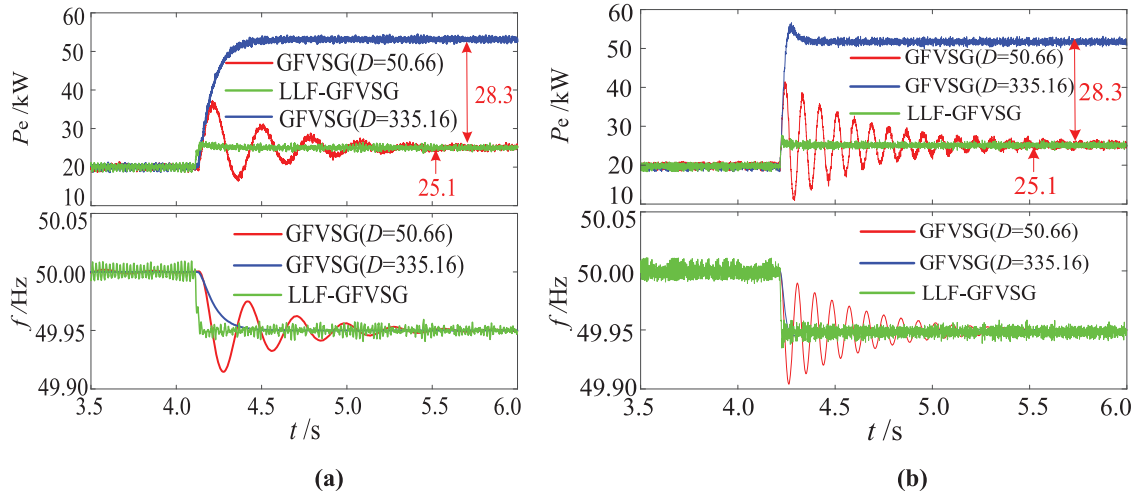


Figure 10: Comparative experimental results under f_g step. (a) $X = 0.1 \Omega$. (b) $X = 0.05 \Omega$

According to the experimental comparison results shown in Figs. 9a and 10a, it is evident that the experimental outcomes during the transitions of P_{ref} from 20 to 60 kW and f_g from 50 to 49.95 Hz maintain a one-to-one correspondence with the simulation results presented in Fig. 5. According to the experimental comparison results shown in Figs. 9b and 10b, it is evident that the system’s damping ratio decreases as the line impedance decreases, resulting in more pronounced oscillations for GFVSG ($D = 50.66$). In contrast, the effectiveness and adaptability of LLF-GFVSG under the same conditions of reduced line impedance are markedly superior to those of GFVSG. Specifically, when the value of D increases from 50.66 to 335.16, the P_e and f for the GFVSG grid-connected system do not vibrate under the P_{ref} and f_g disturbances. However, the P_e has an active steady state deviation of $\Delta P_{e0} = 28.3$ kW at $f_g = 49.95$ Hz, while the P_e and f of LLF-GFVSG do not have dynamic oscillation when responding to two disturbances of P_{ref} and f_g , and the P_e can eliminate the active steady state deviation at $f_g = 49.95$ Hz. So ΔP_{e01} is equal to 0.

In order to verify the effectiveness of LLF-GFVSG, we conducted parallel networking verification on the two GFVSGs shown in Fig. 7. The main system parameters for the parallel network system were configured as follows: $P_{\text{ref}1} = 2P_{\text{ref}2} = 40 \text{ kW}$, $D_1 = 2D_2 = 200$, $J_1 = 2J_2 = 6 \text{ kg}\cdot\text{m}^2$, $K_{d1} = 2K_{d2} = 5.3 \times 10^{-5}$, $X_1 = 2X_2 = 0.1 \ \Omega$, $k_{q1} = 2k_{q2} = 1.4 \times 10^{-4}$. Fig. 11 shows the comparison of experimental results between LLF-GFVSG and traditional GFVSG during the running of two GFVSGs parallel networking. At the initial time, the GFVSG1 and GFVSG2 jointly support a 60 kW resistive load to maintain stable operation. The 60 kW resistive step load is applied at 1 s and removed at 1.6 s. As shown in Fig. 11a, the LLF-GFVSG parallel networking control method can effectively mitigate the dynamic oscillation of the output active power P_e in the parallel networking system under step disturbances of the system load, compared to the existing traditional GFVSG parallel networking control method. Similarly, as shown in Fig. 11b, the LLF-GFVSG parallel networking control method can also effectively address the dynamic oscillation of the output frequency f_s in the parallel networking system under step disturbances of the system load, compared to the existing GFVSG parallel networking control method.

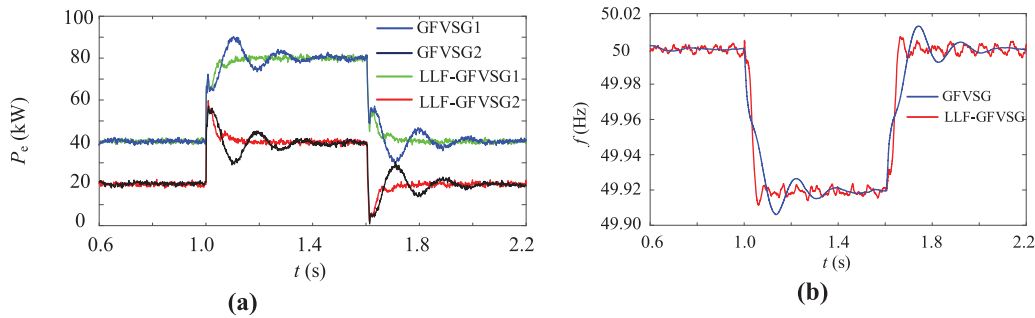


Figure 11: Comparative simulation results under load step. (a) P_{e1} and P_{e2} . (b) f_s

It is worth noting that because the proposed LLF-GFVSG directly employs ΔP for dynamic feed-forward compensation of $\Delta\omega$, the harmonic components present in P_e are introduced into f , resulting in the experimental test waveform of f containing more harmonics compared to that of the GFVSG. Consequently, the dynamic response waveform of f in the LLF-GFVSG exhibits a coarser envelope shape. Therefore, optimizing the dynamic response performance and waveform quality of the f of LLF-GFVSG when dealing with P_{ref} and f_g disturbances is one of the subsequent research works.

5 Conclusion

In order to solve the GCAP dynamic oscillation problem of GFVSG based on the first-order low-pass filter similar to that of TSG, an improved LLF-GFVSG control strategy is proposed. By means of theoretical analysis, mathematical modeling, parameter design, simulation, experimental comparison and verification, the following conclusions are derived:

(1) GFVSG can realize the dynamic oscillation suppression of P_e and f under the two disturbances of P_{ref} and f_g by increasing the value of D , but at the same time, increase the active steady state deviation of P_e under the condition that f_g deviates from the rated value. In other words, the GFVSG has the problem that the GCAP dynamic response performance and its steady state performance cannot be balanced.

(2) The LLF-GFVSG can ensure that there are no dynamic oscillations in P_e and f under variations in parameters U_g , X , Q_{ref} , P_{ref} , and f_g , as well as in a parallel network system. Furthermore,

the steady state deviation of P_e remains zero when f_g deviates from its rated value, indicating that the LLF-GFVSG effectively resolves the challenge of balancing GCAP's dynamic response performance with its steady-state performance, a limitation found in traditional GFVSG.

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Availability of Data and Materials: The authors confirm that the data supporting the findings of this study are available within the article. The additional data that support the findings of this study are available on request from the corresponding author, upon reasonable request.

Ethics Approval: Not applicable.

Conflicts of Interest: The authors declare that they have no conflicts of interest to report regarding the present study.

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